

OUSED in either a cream or a brown plastic cabinet, the Bush DAC90A is a 4-valve (plus rectifier) 2-band transportable superhet designed to operate from A.C. or D.C. mains of 200-250 V, 40-100 c/s in the case of A.C. The waveband ranges are 187-560 m and 1,070-1,900 m.

An earlier version of the DAC90A, model DAC90, is covered in Service Sheet 950 and is easily identified by its tuning scale, which is lighter in colour, and by its waveband control, which is positioned half-way up the cabinet side instead of at its lower edge as in the DAC90A.

Release date and original prices: February 1950, £12 is 8d (brown); £12 is 9d (cream). P.T. extra.

## CIRCUIT DESCRIPTION

Tuned frame aerial inputs L2, C25 (M.W.) and L1, L2, C25 (L.W.) precede triode hexode valve (V1, Wullard UCH42) operating as frequency changer with internal coupling. S1 closes to short-circuit L1 for M.W. operation.

Single oscillator grid coil L3 is tuned by C26 and is used on both wavebands. Parallel trimming by C28 (M.W.) and C8, C27, C28 (L.W.); series tracking by C7 (M.W. and L.W.). Reaction coupling from oscillator anode by L4.

Second valve (V2, Mullard UF41) is a

# BUSH DAC90A

Transportable A.C./D.C. Superhet

variable-mu R.F. pentode operating as intermediate frequency amplifier with tuned transformer couplings C3, L5, L6, C4 and C11, L7, L8, C12.

Intermediate frequency 465 kc/s.

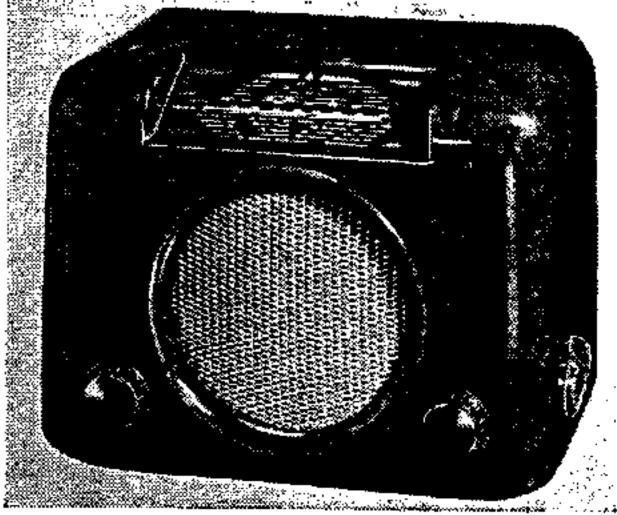
Diode signal detector is part of double diode triode valve (V3, Mullard UBC41).

(Continued col. 1 overleaf)

# COMPONENTS AND VALUES

 	CAPACITORS	Values	Loca- tions
C1	A.G.C. decoupling	0·05µF	G3
C2	L.W. aerial trim	130pF	H4
C3	lat L.F. trans	110pF	<b>B2</b>
C4	tuning {	110pF	B2
C5	V1 cath. by-pass	0.05µF	G3
C6	V1 osc. C.G	50pF	G4
C7	Osc. tracker	605pF	H4
C8	L.W. osc. trim	515pF	H4
C9	A.G.C. decoupling	$0.05\mu F$	F8
C10	H.T. decoupling	$0.05 \mu F$	G3
C11	2nd I.F. trans.	110pF	C2
C12	ftuning {	110pF	C2
C13	I.F. by-pass	100pF	F4
C14	A.G.C. coupling	50pF	F4
C15	A.F. coupling	$0.01 \mu F$	<u>F</u> 3
C16	V3 cath. by-pass	$0.05 \mu F$	F3
C17	Tone corrector	$0.003\mu F$	<u>F</u> 3
C18	A.F. coupling	$0.01 \mu F$	E3
C19	Tone corrector	$6.01\mu\mathbf{F}$	D1
C20*	H.T. smoothing {	$32\mu F$	B1
C21*	)	$16\mu F$	B1
C22	Mains R.F. by-pass	$0.1 \mu F$	D1
C23‡	L.W. aerial trim	40pF	G4
C24‡	M.W. aerial trim.	40pF	G4
C25†	Aerial tuning	528pF	A2
C26†	Oscillator tuning	528pF	A2
C27‡	L.W. osc. trim	40pF	G4
C28‡	M.W. osc. trim	40pF	G4
* <b>E</b> ]	ectrolytic. † Variable.	t Pre-s	et.

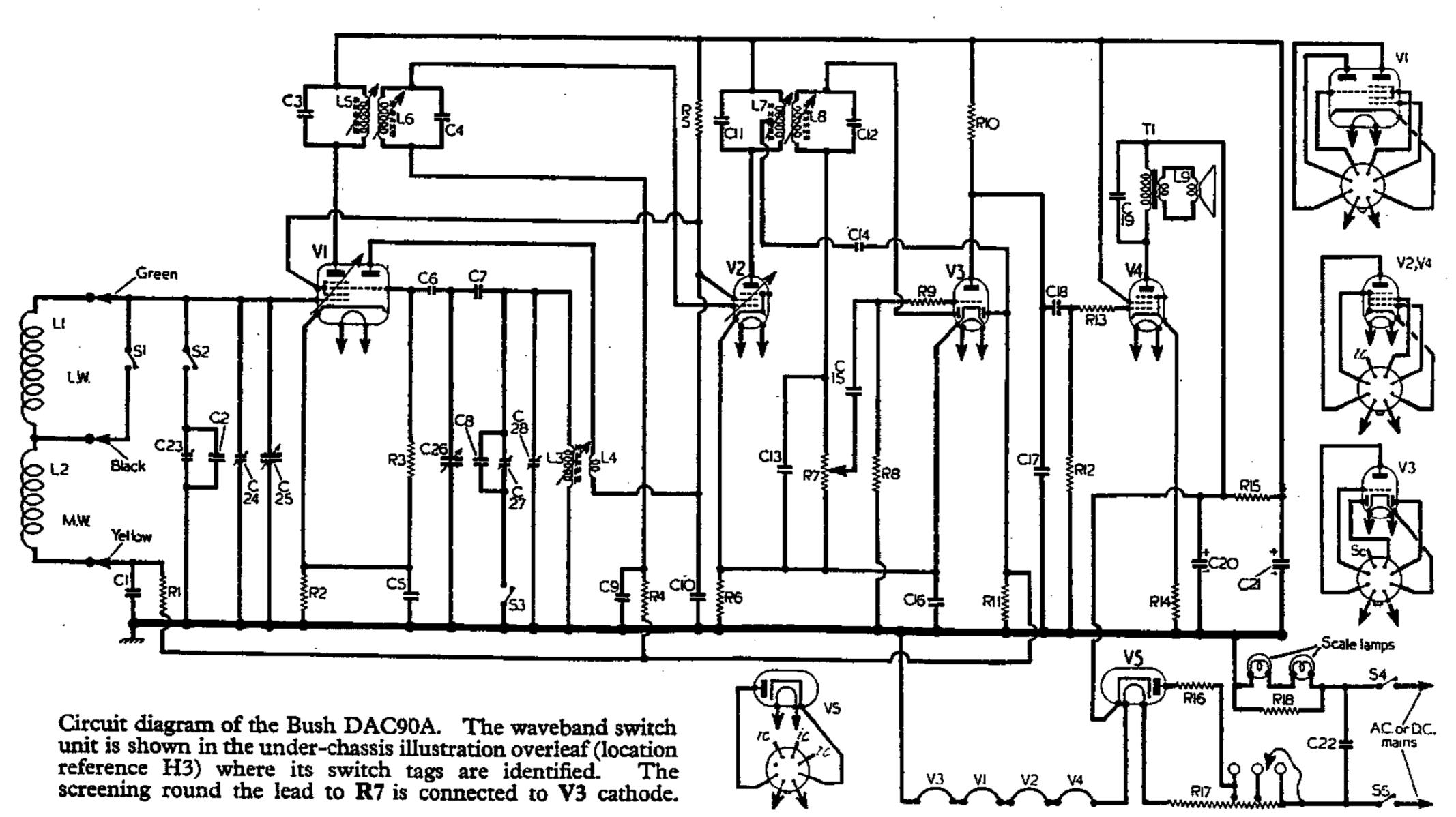
Electrolytic. † Variable. ‡ Pre-set.



Appearance of the Bush DAC90A

RESISTORS		Values	Loca- tions
R1 R2 R3 R4 R5 R6 R7 R8 R10 R11 R12 R13 R14 R15 R16 R17	A.G.C. decoupling V1 G.B. V1 osc. C.G. A.G.C. decoupling H.T. feed V2, V3, G.B. Volume control V3 C.G. V3 C.G. V3 anode load A.G.C. diode load V4 C.G. V4 C.G. stopper V4 G.B. H.T. smoothing V5 surge limiter Heater ballast Scale lamp shunt	Values  1MΩ 220Ω 47kΩ 2.2MΩ 12kΩ 330Ω 500kΩ 2.2MΩ 100kΩ 150kΩ 150kΩ 150kΩ 150Ω 150Ω 150Ω 150Ω 1750Ω 1,250Ω† 75Ω	F3 G3 F3 F3 F4 F4 E4 D2 D1 C2

† Tapped at  $950\Omega + 150\Omega + 150\Omega$ .



OTB	ER COMPONENTS	Approx. Values (ohms)	Loca- tions
L1 L2 L3 L4 L5 L6 L7 L8 L9 T1 S1-S3 S4, S5	L.W. frame aerial M.W. frame aerial Osc. tuning coil Osc. reaction coil  lst I.F. trans. { Pri. Sec. Waveband switches Mains sw., g'd R7		A2 A1 H4 H4 B2 C2 C2 C2 D1

#### Circuit Description—continued

Audio frequency component in its rectified output is developed across volume control R7, which also acts as diode load, and is passed via C15 to grid of triode section.

Second diode of V3, fed via C14 from a tapping on L7, provides a D.C. potential which is developed across R11 and fed back as bias to V1 and V2, giving automatic gain control. Delay voltage, together with bias for V2 and V3, is developed across R6.

Resistance-capacitance coupling by R10, C18 and R12 between V3 and pentode output valve (V4, Mullard UL41). Fixed tone correction by C19 in V4 anode circuit, and by the negative feed-back introduced by omitting the normal by-pass capacitor in V4 cathode circuit.

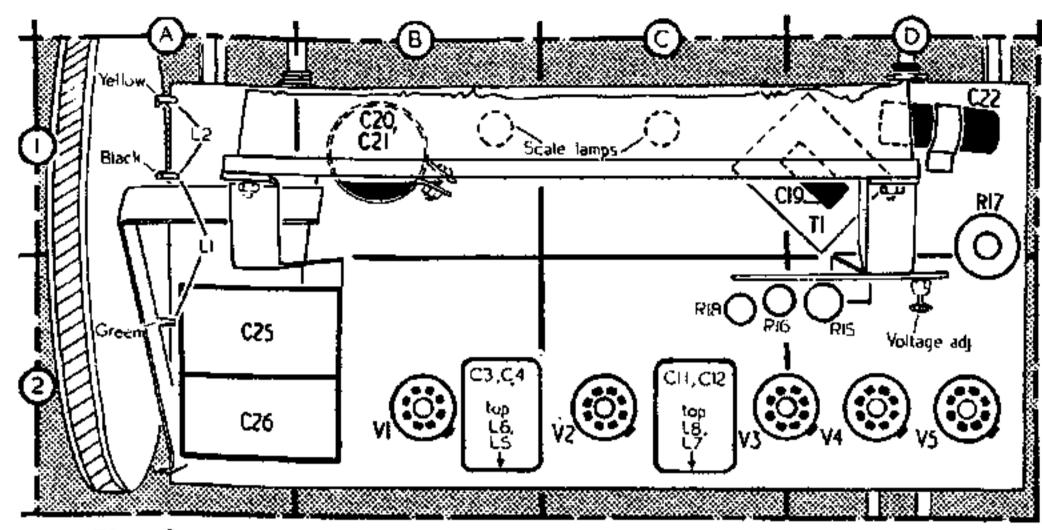
H.T. current is supplied by I.H.C. half-wave rectifying valve (V5, Mullard UY41). Smoothing by R15 and electrolytic capacitors C20, C21. Valve heaters, together with scale lamps and ballast resistor R17, are connected in series across the mains input. Mains R.F. filtering by C22.

#### **GENERAL NOTES**

Switches.—S1-S3 are the waveband switches, ganged in a single rotary unit beneath the chassis. This is shown in detail in our underchassis illustration. In the M.W. position (control knob anti-clockwise) \$1 closes; in the L.W. position, \$2 and \$3 close.

Scale Lamps.—These are rated at 3.5 V, 0.15 A. They have small, clear, spherical bulbs and M.E.S. bases.

Tuning Drive Replacement.—About 31 feet of nylon braided glass yarn is required for a new drive cord which should be run as indicated in the sketch in column 2.

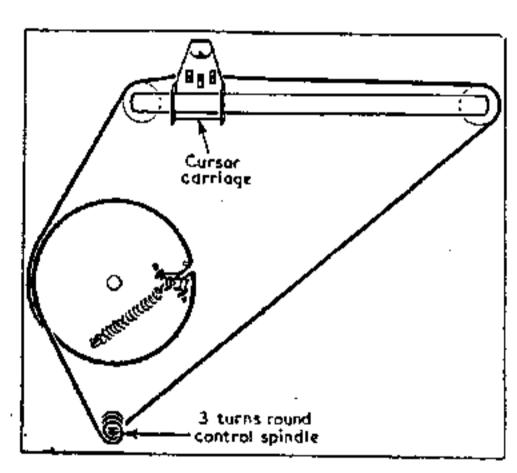


Plan view of the chassis, showing the colour-coded frame aerial connections.

### CIRCUIT ALIGNMENT

Access to all the I.F., aerial and oscillator adjustments used in the following alignment instructions can be gained simply by removing the cabinet back cover.

1.F. Stages.—Screw out the core adjustments of L5, L6, L7 and L8 to their fullest extent. Connect signal generator, via an isolating

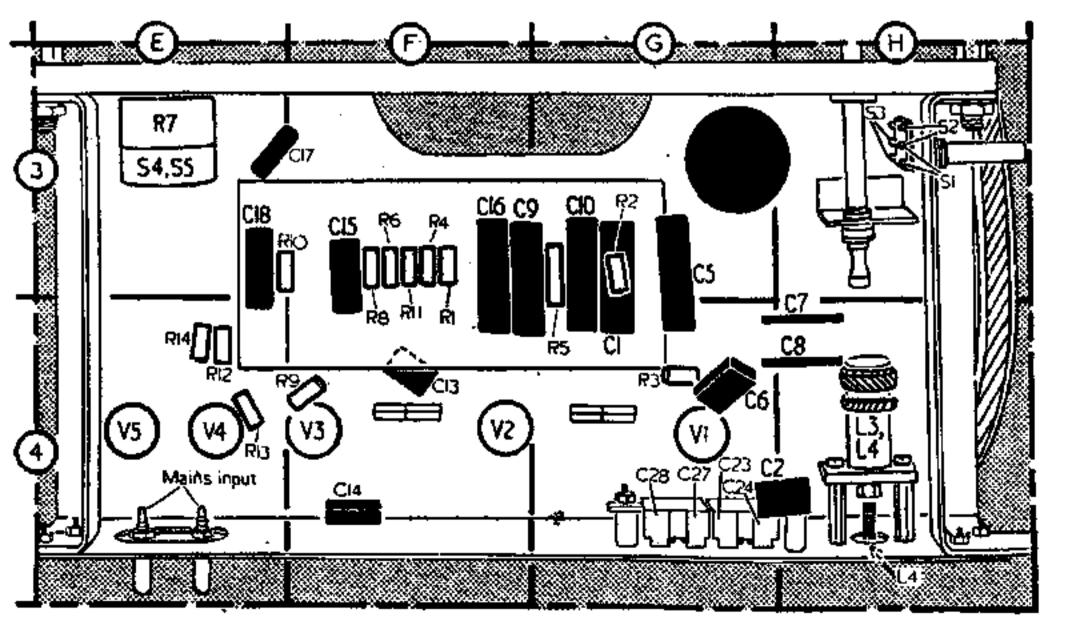


Sketch of the drive cord system drawn as seen from rear of chassis with gang at maximum.

capacitor in each lead, to control grid (pin 6) of V2 and chassis. Switch receiver to M.W. and tune to a point around 300 m which is

maximum.

capacitor in each lead, to control grid (pin 6)



Under-side view of chassis, showing the waveband switch unit in detail (location H3).

clear of powerful signals. Feed in a 465 kc/s (645.16 m) signal and adjust the cores of L8 (location reference C2) and L7 (C2) for maximum output. Transfer "live" signal generator lead to control grid (pin 6) of V1, feed in a 465 kc/s signal and adjust L6 (B2) and L5 (B2) for maximum output. Repeat these adjustments, starting again at L7 and L8 with the signal generator connected to V2 control grid.

R.F. and Oscillator Stages .- If the receiver is to he aligned out of its cabinet, use may be made of the calibration points on the metal scale reflector plate. These points take the form of a line of indentations on the top rear edge of the plate. Viewed from the rear of the chassis and reading from left to right, these indentations represent the following calibration points: Maximum capacitance setting of gang; 500 m; 1,400 m; 800 m; 1,200 m; 200 m. Check that with the gang at maximum capacitance, the cursor coincides with the maximum capacitance calibration point on the reflector plate or with the vertical lines at the high wavelength end of the tuning scale.

The signal generator should be coupled to the receiver via a single loop of wire about the same size as the frame aerial, and placed 12 to 18 inches away from it. The M.W. alignment should be carried out first as 628 and 624 are common to both wavebands and will affect L.W. adjustments.

M.W.—Switch receiver to M.W., tune to 500 m, feed in a 500 m (600 kc/s) signal and adjust the core of L3, L4 (H4) for maximum output. Tune receiver to 200 m, feed in a 200 m (1.500 kc/s) signal and adjust C28 (G4) and C24 (G4) for maximum output. Repeat these adjustments.

L.W.—Switch receiver to L.W., tune to 1,402 m. feed in a 1,402 m (214 ke/s) signal and adjust C27 (G4) and C23 (G4) for maximum output.

# VALVE ANALYSIS

Valve voltages and currents given in the table below are those derived from the manufacturer's information. They were measured on a receiver operating from 230 V A.C. mains, and switched to M.W. There was no signal input.

Voltages were measured on the 1.000 V and 10 V ranges of a Model 7 Avometer, chassis being the negative connection. Total cathode current of the rectifier V5 was quoted at 39 mA.

Valve	Anode		Screen		Cath.
	v	BIA	V	mA	v
VI UCH42	88 Osci 47	1.5 }   listor   0.5	47	1-6	0-8
V2 UF41	98	2.5	47	0.8	1.0
V3 UBC41 V4 UL41	74	0.2	} =		1.0
	190	27.5	98	<b>4</b> ·0	_ 5∙0
V5 UY41	222*	<del>-</del>	ļ —	<del></del>	205.0

A.C. reading.